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## Theory of Optically Controlled Millimeter-Wave Phase Shifters

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**Abstract** — In this paper we analyze the millimeter-wave propagation characteristics of a dielectric waveguide containing a plasma-dominated region. Such a device presents a new method for controlling millimeter-wave propagation in semiconductor waveguides via either optical or electronic means resulting in ultrafast switching and gating. We have calculated the phase shift and attenuation resulting from the presence of the plasma. Higher order modes, both TE and TM, as well as millimeter-wave frequency variation, are studied in both Si and GaAs dielectric waveguides. We have also formulated a surface plasma model that is a good approximation to the more elaborate volume plasma model. Phase shifts are predicted to be as high as  $1400^\circ/\text{cm}$  for modes operating near cutoff. These modes suffer very little attenuation when the plasma region contains a sufficiently high carrier density.

### I. INTRODUCTION

WE ARE CURRENTLY witnessing a resurgence of interest in millimeter-wave technology. The frequency band extending from 30 to 1000 GHz is attractive in several respects. Devices operating above *K*-band frequencies offer greater carrier bandwidth, better spatial resolution, and a more compact technology than presently used *X*- and *K*-band systems. Millimeter- and submillime-

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ter-wave systems also have some advantages over optical systems, such as better atmospheric propagation in selected bands and a technology more amenable to frequency multiplexing [1]–[3], while retaining good angular resolution of the latter. A basic problem is how to effectively preserve these benefits. Our approach to the solution of this problem promises to yield much larger modulation bandwidths than are realizable with optical systems, while preserving the economy of the millimeter-wave system over a given carrier frequency band.

One of the important parts of the microwave and/or millimeter-wave system is the waveguide. At microwave frequencies, metal waveguides are commonly used. At higher frequencies, either microstrip or dielectric waveguide structures become more attractive. Microstrip transmission lines are used up to 30 GHz. For frequencies greater than 30 GHz, the losses in microstrip structures are high, and fabrication techniques become more difficult due to the small strip width and the substrate thickness. Dielectric rectangular waveguides become an alternative to the expensive metal waveguides. The use of high-purity semiconductor materials as dielectric waveguides is particularly important since active devices such as oscillators, Gunn or IMPATT diodes, mixers/detectors, and modulators can be fabricated monolithically with the semiconducting waveguides.

One important aspect of millimeter-wave devices is the control of the phase and amplitude of a wave propagating through the waveguide. The use of semiconductor bulk phenomena in implementing microwave control components has been discussed [4]. The principal phenomena explored have been the dielectric and conductive properties of the plasma state. A frequency-scanning millimeter-wave antenna utilizing periodic metallic-stripe perturbations on a silicon waveguide has been demonstrated [5]. Millimeter-wave dielectric image-guide integrated devices have been developed [6]. In the optical region, there are a variety of controllable waveguide devices which have not found their counterparts in the millimeter-wave region. High-speed light modulators that make use of the electrooptic, acoustooptic, and magnetooptic effects in bulk material have been described [7].

Phase shifting is a fundamental control operation. A general approach to this operation, used extensively at both optical [8] and microwave [9] frequencies, is to alter the phase velocity along a fixed interval of a guiding medium. In this case, the phase shift per unit length is equal to the change in the propagation constant of the guided wave. One well-known method of altering the dispersion of millimeter waves is to introduce a plasma into the guiding medium. In their work, Jacobs *et al.* [10] have demonstrated that the phase shift can be accomplished by injecting plasma with p-i-n diodes [11]. Similar work at low frequency has been demonstrated recently by Glance [12]. There are several shortcomings of p-i-n-diode-controlled millimeter-wave devices: a) large phase shifts have not been achieved due to excess heating of the waveguide; b) there are large losses ( $> 6\text{dB}$ ) due to excess metallization for contact to the junction region; and c) the p-i-n diode becomes an integral part of the waveguide. This leads to complex boundary conditions and poor electrical isolation between the p-i-n diode and the waveguide structure.

Recently, we have demonstrated a way of circumventing the above difficulties. By illuminating the guide with above-bandgap radiation, we have created the basic element of a new class of device, an optically controlled millimeter-wave phase shifter and modulator. Optical control offers the following advantages: a) near perfect isolation; b) low static and dynamic insertion losses in some regimes; c) fast response; d) high power handling capability; e) when picosecond pulses are used, it is possible for extremely high density plasmas to be injected without damaging the material; f) by proper choice of semiconductor material and laser wavelength, one can generate plasma with any desirable density distributions and at any desirable time; g) ultrafast switching and gating of millimeter-wave signals is possible; and h) using picosecond exciting-probing techniques, the dynamic evolution of the injected plasma can be studied in detail. Parameters related to transport properties of the carriers can be accurately determined.

This paper analyzes in detail the changes that occur in the propagation characteristics of millimeter waves in a

dielectric waveguide when a plasma-dominated region is present. The model and method analysis are presented in [13]. In that paper, both the experimental and theoretical studies concentrated primarily on the propagation of the dominant lowest order mode (TM) denoted by  $E_{1,1}^y$  for 94-GHz waves in a silicon waveguide. In practice, it has been found that for waves propagating in an over-sized waveguide with a plasma covered region, the waves will not remain in the fundamental mode. In this paper, we extend the analysis to include higher order TM modes  $E_{p,q}^y$  and TE modes  $E_{p,q}^x$ . Calculations have also been carried out at frequencies other than 94 GHz. To understand the multi-mode propagation characteristics more fully, a model based on a surface plasma is devised, and the results compared with those obtained by a more elaborate calculation involving a finite thickness plasma layer. It is found that the surface plasma model gives a good approximation for plasma layers less than  $10\text{ }\mu\text{m}$  in thickness. This is significant since we now have a simple theoretical model to predict the phase shift and to understand the mode-coupling mechanism in a general situation for plasma-controlled phase-shifting devices of arbitrary dimensions.

Recently, we have demonstrated the operation of a wide-bandwidth, high-repetition rate opto-electronic modulator for 94-GHz signals based on a plasma-controlled Cr-doped GaAs waveguide [14]. The rapid phase and amplitude modulation are achieved by using picosecond optical pulses to inject a plasma into the waveguide. In Cr-doped GaAs, in contrast to silicon, the electron-hole plasma recombines rapidly, i.e., in less than 100 ps, thus resulting in a high-speed, wide-bandwidth, and high-repetition rate operation of a millimeter-wave modulator. Therefore, in this paper, we extend our theoretical investigation of dielectric materials to include GaAs.

In Section II, the concept of a plasma-controlled dielectric waveguide is reviewed and earlier experimental results are summarized. The complete analysis, including the presentation of the surface plasma model, is then treated in Section III. It should be pointed out that the analysis presented here is quite general; it can be applied to the plasma-controlled dielectric waveguide devices no matter whether the plasma injection is by optical means, as in this work, or by electronic means, as in the work reported by Jacobs *et al.* [10].

## II. CONCEPT OF PLASMA-CONTROLLED DIELECTRIC WAVEGUIDE WITH PRELIMINARY EXPERIMENTAL RESULTS

Shown in Fig. 1 is a schematic of the optically controllable phase shifter. It consists of a rectangular semiconductor waveguide with tapered ends to allow efficient transition of millimeter waves both to and from a conventional metal waveguide. Optical control is realized when the broadwall of the semiconductor guide is illuminated by light generated by either a proximal source or one removed by a suitable optical guiding medium. The width  $a$  and height  $b$  of the guide are selected so that it supports an  $E_{1,1}^y$  mode.

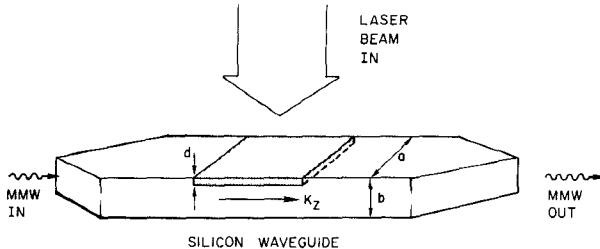


Fig. 1. Schematic diagram of the optically controlled phase shifter. The propagation vector in the guide is  $k_z$ ,  $d$  is the depth of the injected plasma layer,  $a$  is the width of the guide, and  $b$  is the height of the guide.

The initial depth of plasma injection is controlled by selecting an appropriate combination of optical radiation wavelength and semiconductor absorption properties.

At sufficiently small injection depths, the final thickness of the plasma is determined primarily by processes of carrier diffusion and recombination. The effect of the plasma-occupied volume is to introduce a layer whose index of refraction at the millimeter-wave frequency is larger than that of the remaining volume of the waveguide. As the optical-illumination intensity increases from a low value, a significant phase shift will not appear until the plasma frequency exceeds the frequency of the guided millimeter wave, after which it will rapidly rise and eventually saturate. The form and magnitude of the phase shift versus the intensity of illumination, and hence the plasma density, depends in detail on the material and geometrical factors that characterize the optically perturbed guiding structure, and the determination of these quantities requires a detailed solution of the corresponding boundary value problems. The optically induced phase shift  $\Delta\phi$  for a given section of waveguide of length  $l$  is determined by computing the propagation constant in the  $z$  direction; first, in the absence of the injected plasma  $k_z$ , and then with the plasma  $k'_z$ ; thus,  $\Delta\phi = (k'_z - k_z)l$ .

Comparison between the experimental and the theoretical results presented in [13] and in Section III are shown in Fig. 2. The correspondence between the data and the theoretically predicted curve is excellent. As shown in Fig. 2, we observed a maximum phase shift of  $59^\circ$  at 94 GHz for a plasma column 1.6 mm in length with dynamic insertion loss of less than one decibel.

By slightly variations of the phase-shifting technique, we have also demonstrated opto-electronic switching and gating of millimeter-wave signals at 94 GHz by an optically induced plasma in a silicon waveguide [15]. A millimeter-wave pulse width as short as one nanosecond and variable to tens of nanoseconds can readily be obtained by this technique. We conclude from these studies that optical injection of plasma is superior to electrical injection because of its near-perfect isolation. The geometry and nature of the injected-plasma column are more amenable to analysis. Excellent agreement between experiment and theory encourages further study and a search for better experimental configurations to realize a new class of devices with high efficiency.

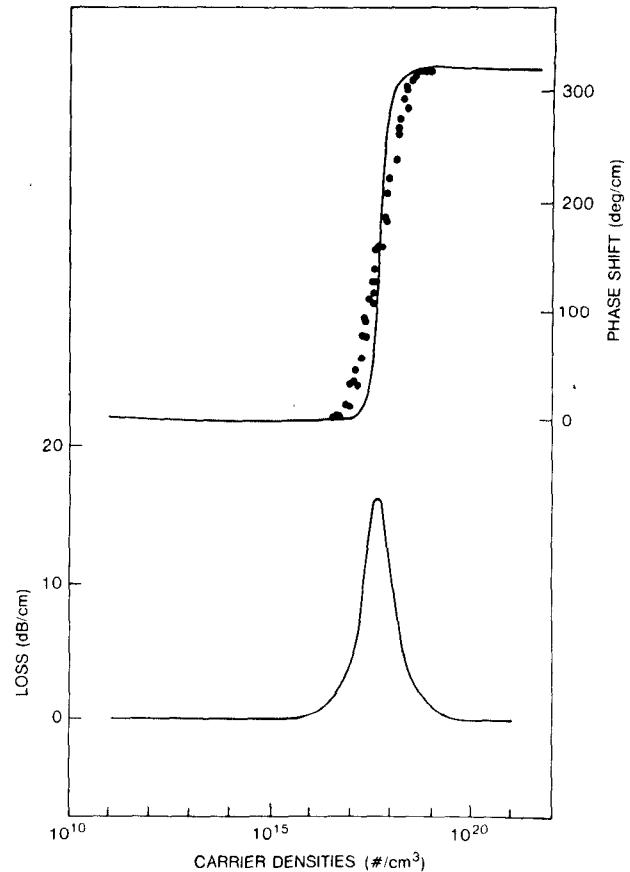


Fig. 2. Measured phase shift normalized to units of deg/cm [15].

### III. MODEL AND ANALYSIS OF A PLASMA-CONTROLLED DIELECTRIC WAVEGUIDE

The geometry of the rectangular dielectric-plasma guide model is shown in Fig. 3. The guide cross section is in the  $x-y$  plane, and the  $z$  coordinate represents the direction of propagation. The dimensions of the guide are denoted by  $a$  and  $b$ , and the plasma region thickness by  $t_p$ . The medium surrounding the waveguide is air. The relative dielectric constant and refractive index of the dielectric region  $0 < y < b - t_p$  are  $\epsilon_r$  and  $n_r = \sqrt{\epsilon_r}$ , respectively, and those of the dielectric-plasma region  $b - t_p < y < b$  are [13]

$$\epsilon_p = n_p^2 = \epsilon_r - \alpha \frac{\omega_{p\alpha}^2}{\omega^2 + \nu_\alpha^2} - j\alpha \frac{\omega_{p\alpha}^2}{\omega^2 + \nu_\alpha^2} \frac{\nu_\alpha}{\omega} \equiv (\eta - j\kappa)^2. \quad (1)$$

The free-charge contribution (plasma) is characterized by a plasma frequency  $\omega_{p\alpha} = (q_\alpha^2 N_\alpha / m_\alpha \epsilon_0)^{1/2}$  of each species of density  $N_\alpha$ , and an effective collision frequency  $\nu_\alpha$ . These various species in an optically formed plasma are denoted by thermally ionized and photo-induced holes (light and heavy) and electrons. The values of the various species properties, that is, density, effective mass, and collision frequency, are listed in Fig. 4(a) for Si and Fig. 4(b) for GaAs. In Fig. 4(a) the real part  $\eta$  and imaginary part  $\kappa$  of the refractive index for Si (see (1)) are plotted as a function

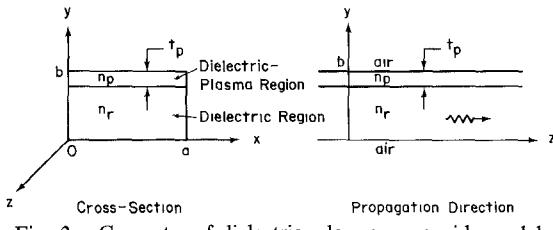


Fig. 3. Geometry of dielectric-plasma waveguide model.

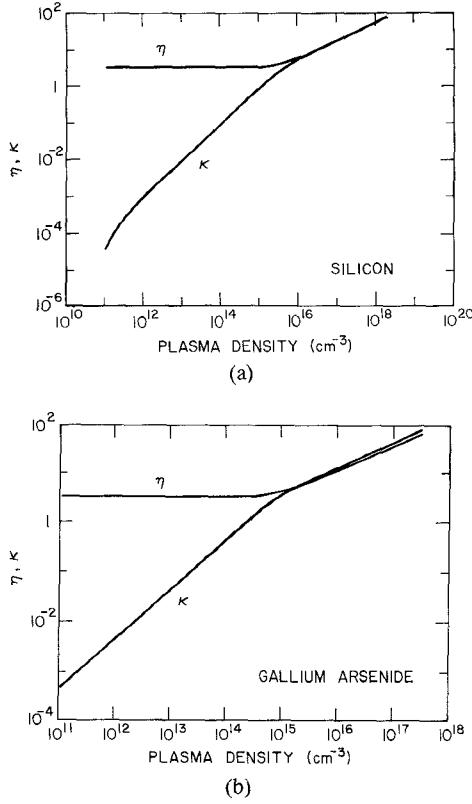


Fig. 4. (a) Plot of the plasma region refractive index versus plasma density in Si at 94 GHz. Here  $P_L = 0.14 N_p$ ;  $P_H = 0.86 N_p$ ;  $m_e^* = 0.259 m_0$ ;  $m_p^* = 0.38 m_0$ ;  $m_p^* = 0.16 m_0$ ;  $m_p^* = 0.49 m_0$ ;  $\mu_e = 1500 \text{ cm}^2/\text{V}\cdot\text{s}$ ;  $\tau_e = 2.2 \times 10^{-13} \text{ s}$ ;  $\mu_h = 600 \text{ cm}^2/\text{V}\cdot\text{s}$ ;  $\tau_p = 1.3 \times 10^{-13} \text{ s}$ ; and  $N_{p0} = 10^{11} \text{ cm}^{-3}$  [16], [17]. (b) Plot of the plasma-region refractive index versus plasma density in GaAs at 94 GHz. Here  $P_L = 0$ ;  $P_H = N_p$ ;  $m_e^* = 0.06 m_0$ ;  $m_p^* = 0.5 m_0$ ;  $m_p^* = 0.8 m_0$ ;  $\mu_e = 8800 \text{ cm}^2/\text{V}\cdot\text{s}$ ;  $\tau_e = 3.3 \times 10^{-13} \text{ s}$ ;  $\mu_h = 450 \text{ cm}^2/\text{V}\cdot\text{s}$ ;  $\tau_p = 1.28 \times 10^{-13} \text{ s}$ ; and  $N_{p0} = 1.4 \times 10^6 \text{ cm}^{-3}$  [18-20].

of plasma density for a 94-GHz wave. Likewise, in Fig. 4(b) the refractive index in the plasma-dielectric region for GaAs is plotted. For each material, note the metallic nature of the plasma region for plasma density above about  $10^{16} \text{ cm}^{-3}$  for Si and  $10^{15} \text{ cm}^{-3}$  for GaAs.

For the case of well-guided modes, the analysis is considerably simplified in that the dispersion equation is essentially decoupled between the  $x$  and  $y$  directions [21]. That is, if the fields in the dielectric are assumed to vary as  $\exp i(\omega t - k_x x - k_y y - k_z z)$ , we find that boundary conditions in  $x$  determine  $k_x$ , those in  $y$  determine  $k_y$ , and from these one can then determine  $k_z$ . With these assumptions, the modes decouple into TM, called  $E_{p,q}^y$ , and TE, called  $E_{p,q}^x$  modes. The propagation constant  $k_z$  is computed for the cases with and without the presence of the

plasma region,  $k'_z$  (complex),  $k_z$  (real), respectively, and from these the plasma induced phase shift  $\Delta\phi$  for a given length of waveguide  $l$  can be computed by

$$\Delta\phi = (\text{Re } k'_z - k_z)l, \text{ rad} \quad (2)$$

and the attenuation coefficient by

$$\alpha = \text{Im } k'_z, \text{ cm}^{-1}. \quad (3)$$

Specifically, in a waveguide of dimensions  $a$  and  $b$  and with no plasma present,  $n_p = n_r$  ( $\omega_{p\alpha} = 0$ ),  $k_x$  and  $k_y$  for the  $E_{p,q}^y$  mode are approximated by [21]

$$k_{x_{\text{TM}}} \approx \frac{p\pi}{a} \left[ 1 + \frac{1}{\pi a} \frac{\lambda_0}{(\epsilon_r - 1)^{1/2}} \right]^{-1} \quad (4)$$

and

$$k_{y_{\text{TM}}} \approx \frac{q\pi}{b} \left[ 1 + \frac{1}{\pi b} \right] \frac{\lambda_0}{(\epsilon_r - 1)^{1/2}}^{-1}. \quad (5)$$

Here  $p$  is the number of extrema in the  $x$  direction,  $q$  is the number of extrema in the  $y$  direction, and  $\lambda_0$  is the free-space wavelength of the propagating wave. Similarly, for well-guided  $E_{p,q}^x$  modes,  $k_x$  and  $k_y$  are approximated by [21]

$$k_{x_{\text{TE}}} \approx \frac{p\pi}{a} \left[ 1 + \frac{1}{\pi \epsilon_r a} \frac{\lambda_0}{(\epsilon_r - 1)^{1/2}} \right]^{-1} \quad (6)$$

and

$$k_{y_{\text{TE}}} \approx \frac{q\pi}{b} \left[ 1 + \frac{1}{\pi b} \frac{\lambda_0}{(\epsilon_r - 1)^{1/2}} \right]^{-1}. \quad (7)$$

With these values of  $k_x$  and  $k_y$ , we can compute  $k_z$  in the guide without the plasma present

$$k_z = \left( n_r^2 \frac{\omega^2}{c^2} - k_x^2 - k_y^2 \right)^{1/2} \quad (8)$$

where  $k_x$  and  $k_y$  are given by either their TM values (4) and (5), or their TE values (6) and (7).

Now consider the case when the plasma layer is present, as shown in Fig. 3. Again, assuming the decoupling due to well-guided modes is valid, we note that the solutions for  $k_x$  for the  $E_{p,q}^y$  and  $E_{p,q}^x$  modes are given by (4) and (6). In the  $y$  direction, the solution for  $k_y$  for TM( $E_{p,q}^y$ ) waves is found by solving [13]

$$\tan^{-1} \frac{K_a \epsilon_r}{k_r} + \tan^{-1} \left\{ \frac{k_p}{k_r} \frac{\epsilon_r}{\epsilon_p} \tan \left[ \tan^{-1} \left( \frac{K_a \epsilon_p}{k_p} \right) - k_p t_p \right] \right\} - k_r (b - t_p) + (q-1)\pi = 0 \quad (9)$$

and for TE( $E_{p,q}^x$ ) waves

$$\tan^{-1} \frac{K_a}{k_r} + \tan^{-1} \left\{ \frac{k_p}{k_r} \tan \left[ \tan^{-1} \left( \frac{K_a}{k_p} \right) - k_p t_p \right] \right\} - k_r (b - t_p) + (q-1)\pi = 0. \quad (10)$$

In (9) and (10), we have defined the various  $k_y$ 's in each

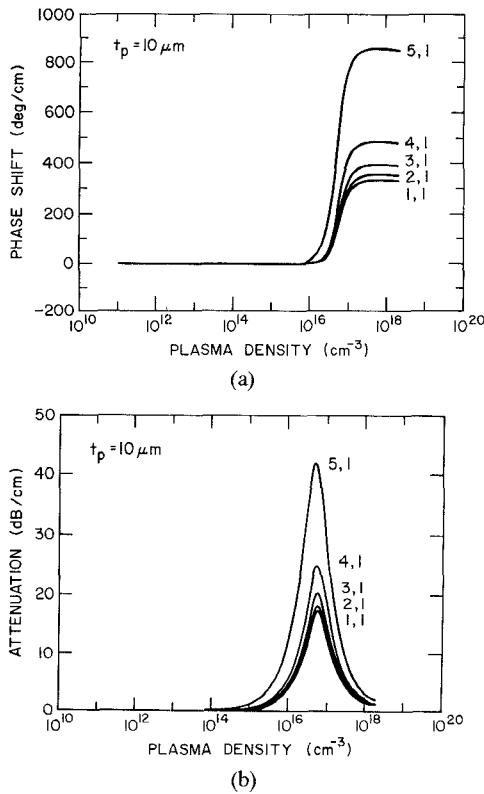


Fig. 5. (a) Phase-shift properties in a  $2.4 \times 1\text{-mm}^2$  Si waveguide at 94 GHz with respect to plasma density for a plasma region thickness of 10  $\mu\text{m}$ . Parametric dependence is the TM mode  $E_{p,q}^y$ . (b) Attenuation properties in a  $2.4 \times 1\text{-mm}^2$  Si waveguide at 94 GHz with respect to plasma density for plasma region thickness of 10  $\mu\text{m}$ . Parametric dependence is the TM mode  $E_{p,q}^y$ .

region (Fig. 3) as

$$\begin{aligned} K_a &= \frac{2\pi}{\lambda_0} (n_\omega^2 - 1)^{1/2} \\ k_r &= \frac{2\pi}{\lambda_0} (n_r^2 - n_\omega^2)^{1/2} \\ k_p &= \frac{2\pi}{\lambda_0} (n_p^2 - n_\omega^2)^{1/2} \end{aligned} \quad (11)$$

and  $n_\omega$  is the effective refractive index of the wave in the decoupled guide. In the plasma layer, the index  $n_p$  is complex and depends upon the density of the plasma (see (1)). Substituting the  $k_y$ 's into (9) and (10) results in transcendental equations for  $n_\omega$ . These equations are solved numerically for  $n_\omega$ , from which  $k_r$  is determined. The value of  $k_r$  is the value of  $k_y$  in the dielectric region in the presence of the plasma layer and thus  $k'_z$  may be found by

$$k'_z = \left( n_r^2 \frac{\omega^2}{c^2} - k_x^2 - k_r^2 \right)^{1/2}. \quad (12)$$

We can then compute the phase shift  $\Delta\phi$  and attenuation  $\alpha$  from (2) and (3), using (8) and (12).

In [15], plots of phase shift and attenuation versus plasma density, with plasma region thickness a parameter, were shown for the lowest order  $E^y$  mode, i.e.,  $E_{1,1}^y$ . The Si waveguide cross sections for those plots were  $2.4 \times 1 \text{ mm}^2$  and  $1 \times 0.5 \text{ mm}^2$ , and the plasma thickness was varied from

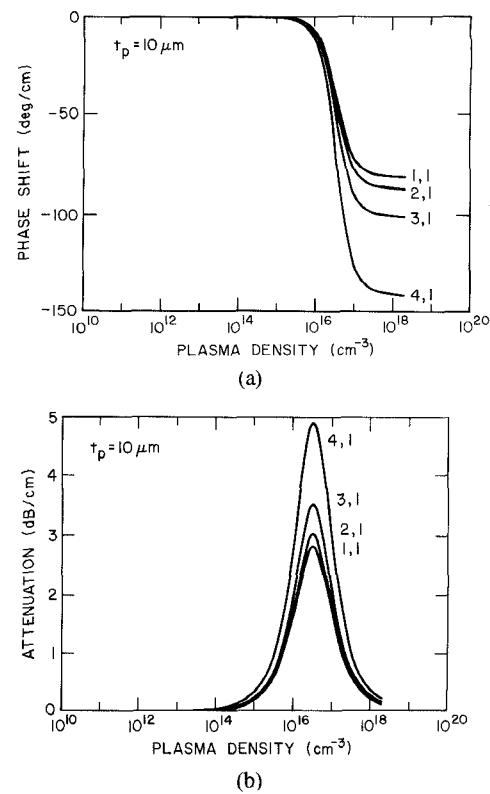
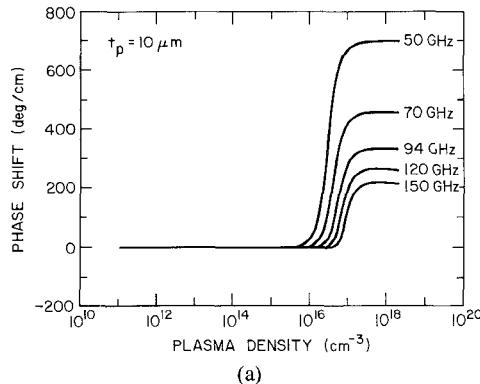
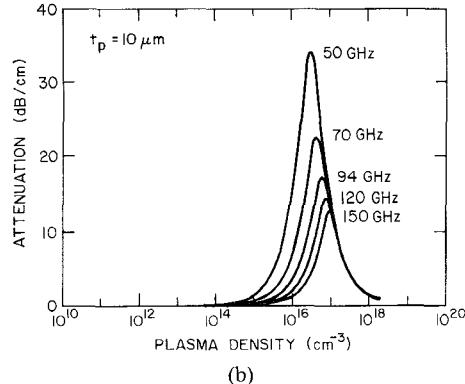


Fig. 6. (a) Phase shift properties in a  $2.4 \times 1\text{-mm}^2$  Si waveguide at 94 GHz with respect to plasma density for a plasma region thickness of 10  $\mu\text{m}$ . Parametric dependence is the TE mode  $E_{p,q}^x$ . (b) Attenuation properties in a  $2.4 \times 1\text{-mm}^2$  Si waveguide at 94 GHz with respect to plasma density for a plasma region thickness of 10  $\mu\text{m}$ . Parametric dependence is the TE mode  $E_{p,q}^x$ .

1  $\mu\text{m}$  to half the guide depth. In this paper, we have calculated  $\Delta\phi$  and  $\alpha$  as a function of the plasma density for a rectangular Si waveguide of dimensions  $2.4 \times 1 \text{ mm}^2$  for all  $E_{p,q}^y$  and  $E_{p,q}^x$  modes that can propagate for a plasma region thickness of 10  $\mu\text{m}$ . The results of the calculations are shown in Figs. 5 and 6. In Fig. 5(a) the phase shift per centimeter is plotted as a function of the plasma density for the  $E_{1,1}^y$  through the  $E_{5,1}^y$  modes, which are the only  $E_{p,1}^y$  modes that can propagate in this guide. The corresponding loss in dB/cm is plotted in Fig. 5(b). As the plasma density increases from  $10^{15} \text{ cm}^{-3}$  to  $10^{20} \text{ cm}^{-3}$ , the skin depth in Si decreases from more than 200  $\mu\text{m}$  to less than 1  $\mu\text{m}$ . When the skin depth is equal to or larger than the plasma layer thickness, the millimeter wave penetrates deeply into the plasma layer causing loss. The maximum loss occurs when the skin depth is about equal to the layer thickness. The higher order modes have more loss in this regime than the lower order modes, because more of the wave power is concentrated in the plasma region when a higher order mode is propagating in the guide. As the plasma density increases further, the skin depth decreases. When the skin depth is less than the thickness of the plasma layer, the plasma region begins to act as a metallic conductor and the dielectric waveguide becomes an image line. The attenuation then drops off rapidly with increasing plasma density. The maximum phase shift of the higher order modes is also larger than that of the lower order modes because the



(a)



(b)

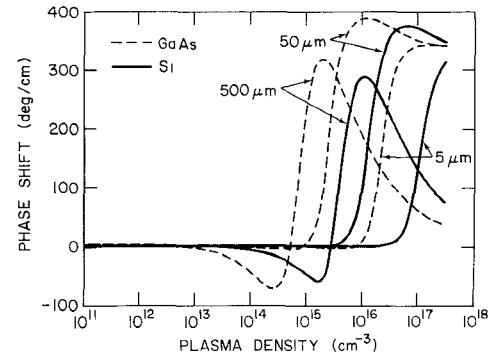
Fig. 7 (a) Phase-shift properties in a  $2.4 \times 1\text{-mm}^2$  Si waveguide with respect to plasma density for a plasma-region thickness of  $10 \mu\text{m}$  and the lowest order TM mode  $E_{1,1}^y$ . Parametric dependence is the frequency of the millimeter wave. (b) Attenuation properties in a  $2.4 \times 1\text{-mm}^2$  Si waveguide with respect to plasma density for a plasma-region thickness of  $10 \mu\text{m}$  and the lowest order TM mode  $E_{1,1}^y$ . Parametric dependence is the frequency of the millimeter wave.

waveguide becomes more dispersive closer to cutoff. The other two TM modes this guide is capable of supporting at 94 GHz are the  $E_{1,2}^y$  and  $E_{2,2}^y$  modes, both of which are of higher order than the  $E_{1,1}^y$  mode. Their maximum phase shifts are  $1000^\circ/\text{cm}$  and  $1450^\circ/\text{cm}$ , respectively.

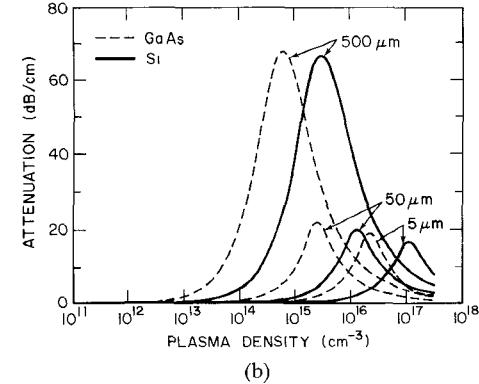
Plots of the phase shift and attenuation curves for the TE modes,  $E_{1,1}^x$  through  $E_{4,1}^x$ , in the  $2.4 \times 1\text{-mm}^2$  guide at 94 GHz are shown in Fig. 6(a) and (b) for the  $10\text{-}\mu\text{m}$  plasma layer. The results are similar to those found previously for the  $E_{p,q}^y$  modes, except that the magnitude of the maximum phase shift is less and the maximum loss in the guide is less, reflecting the different field distribution requirements between the TE and TM modes.

The effect of varying the frequency of the millimeter wave in the  $2.4 \times 1\text{-mm}^2$  guide for a  $10\text{-}\mu\text{m}$  plasma depth is to shift the attenuation peak and the onset of the phase shift. This is shown in Fig. 7(a) and (b) for the  $E_{1,1}^y$  wave. This is to be expected since the skin depth is a decreasing function of frequency; therefore, the plasma density at which the greatest amount of interaction between the plasma and the wave occurs increases as frequency is increased. Also, as the frequency is lowered, the maximum phase shifts and attenuations are larger. This is because the guide operates closer to cutoff at the lower frequencies. For a frequency of 50 GHz, only the mode shown,  $E_{1,1}^y$ , can propagate in the waveguide.

In Fig. 8(a) and (b), we have plotted the results for a



(a)



(b)

Fig. 8. (a) Comparison of the phase-shift characteristics of an  $E_{1,1}^y$  mode propagating at 94 GHz in a  $2.4 \times 1\text{-mm}^2$  waveguide for Si (solid lines) and GaAs (dashed lines) plotted with respect to plasma density. The three sets of curves correspond to plasma depths of  $500 \mu\text{m}$ ,  $50 \mu\text{m}$ , and  $5 \mu\text{m}$ , respectively. (b) Comparison of the attenuation characteristics of an  $E_{1,1}^y$  mode propagating at 94 GHz in a  $2.4 \times 1\text{-mm}^2$  waveguide for Si (solid lines) and GaAs (dashed lines) plotted with respect to plasma density. The three sets of curves correspond to plasma depths of  $500 \mu\text{m}$ ,  $50 \mu\text{m}$ , and  $5 \mu\text{m}$ , respectively.

$2.4 \times 1\text{-mm}^2$  GaAs waveguide. Because the features with respect to multimode and frequency variation are similar to those presented for Si, we have displayed only the lowest order TM mode  $E_{1,1}^y$  results at 94 GHz with plasma thickness a parameter. For comparison, the Si guide results are displayed as solid curves. The general features of the GaAs curves are the same as those for Si; however, the curves are shifted toward lower plasma densities consistent with the shift in the dielectric properties of GaAs versus Si shown in Fig. 4.

#### A. Surface Model of the Plasma

A means of simplifying the calculation of the propagation constant  $k_r$  in the presence of the plasma is to assume that the plasma is located only at the surface with a surface plasma density

$$N_{s\alpha} = N_\alpha t_p, \text{ cm}^{-2} \quad (13)$$

where  $N_\alpha$  is the volume density of each species  $\alpha$ , and  $t_p$  is the thickness of the plasma. With this simplification, the plasma effect only appears in the boundary condition at  $y = b$ . A picture of this simplified model is shown in Fig. 9. The analysis proceeds as before by assuming that the solutions for  $k_x$  and  $k_y$  are decoupled. Again we find  $k_x$  is given by (4) and (6). We examine the plasma's reaction to

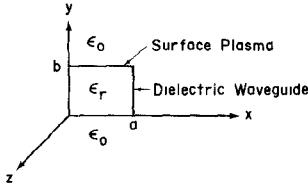


Fig. 9. Cross section of the plasma controlled dielectric waveguide with the plasma treated as a surface density.

the EM fields in terms of the linearized charge density  $\rho$  and a linearized current density  $\vec{J}$ , where

$$\begin{aligned}\rho &= \sum_{\alpha} q_{\alpha} n_{\alpha} \\ \vec{J} &= \sum_{\alpha} q_{\alpha} N_{\alpha} \vec{v}_{\alpha}.\end{aligned}\quad (14)$$

Please note that  $N_{\alpha}$  is the zeroth-order density of species  $\alpha$ , and  $n_{\alpha}$  represents the linearized response to the millimeter wave. Likewise, since the zeroth-order velocity of each species  $\alpha$  is zero,  $\vec{v}_{\alpha}$  represents the linearized velocity response to the millimeter wave. We use the integral form of Maxwell's Equations to find the boundary condition at  $y = b$ . The plasma quantities  $n_{\alpha}$  and  $\vec{v}_{\alpha}$  are found from the linearized fluid equations to be

$$n_{\alpha} = \frac{q_{\alpha}}{m_{\alpha}\omega} \frac{1}{\omega - i\nu_{\alpha}} \left\{ n_{\alpha 0} \left[ \frac{\partial E_x}{\partial x} + \frac{\partial E_y}{\partial y} - ik_z E_z \right] + E_y \frac{\partial n_{\alpha 0}}{\partial y} \right\} \quad (15)$$

and

$$\vec{v}_{\alpha} = \frac{-iq_{\alpha}}{m_{\alpha}} \frac{\vec{E}}{\omega - i\nu_{\alpha}}. \quad (16)$$

We have defined  $n_{\alpha 0} = N_{\alpha} \delta(y - b)$ , that is, a plasma located only at the surface  $y = b$ .

For TM waves, the surface plasma acts as a surface current and the transcendental equation for  $k_r$  becomes

$$\tan^{-1} \epsilon_r \frac{K_a}{k_r} + \tan^{-1} \left\{ \epsilon_r \frac{K_a}{k_r} \left[ \frac{1}{1 - K_a t_p (\epsilon_p - \epsilon_r)} \right] \right\} - k_r b + (q-1)\pi = 0 \quad (17)$$

where  $K_a$  and  $k_r$  are as in (11). For TE waves, the surface plasma acts as a surface charge and the transcendental equation for  $k_r$  becomes

$$\tan^{-1} \frac{K_a}{k_r} + \tan^{-1} \left\{ \frac{K_a}{k_r} \left[ 1 + \left( \frac{2\pi}{\lambda_0} \right)^2 \frac{t_p}{K_a} (\epsilon_p - \epsilon_r) \right] \right\} - k_r b + (q-1)\pi = 0. \quad (18)$$

While (17) and (18) are still transcendental, their solutions are much simpler than those of (9) and (10). Fig. 10(a) and (b) shows the results of the surface plasma model analysis in comparison to the volume model presented in the previous figures. Good agreement of the phase shifts and attenuations of the two models with respect to plasma density at small plasma depth for a TM wave at 94 GHz is seen. As the thickness of the plasma layer gets larger, the

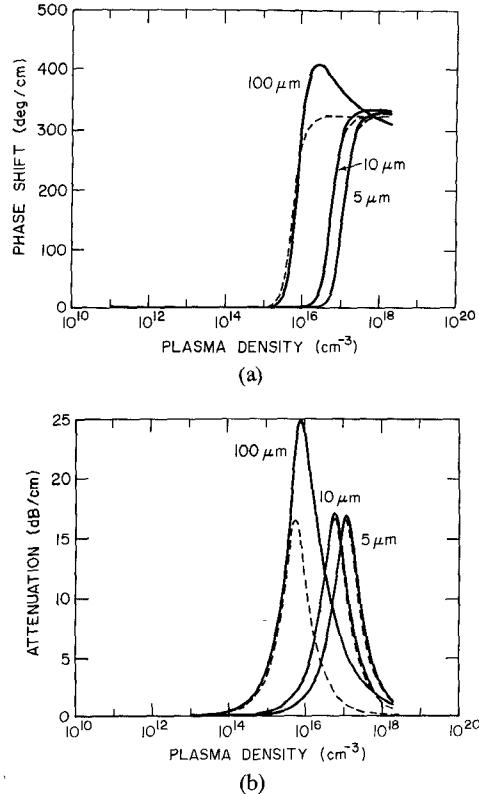


Fig. 10. (a) Comparison of the phase-shift characteristics of an  $E_{1,1}^r$  mode propagating at 94 GHz in a  $2.4 \times 1\text{-mm}^2$  Si waveguide for the surface model (dashed lines) and the volume model (solid lines) plotted with respect to plasma density. The three sets of curves correspond to plasma depths of  $100 \mu\text{m}$ ,  $10 \mu\text{m}$ , and  $5 \mu\text{m}$ , respectively. (b) Comparison of the attenuation characteristics of an  $E_{1,1}^r$  mode propagating at 94 GHz in a  $2.4 \times 1\text{-mm}^2$  Si waveguide for the surface model (dashed lines) and the volume model (solid lines) plotted with respect to plasma density. The three sets of curves correspond to plasma depths of  $100 \mu\text{m}$ ,  $10 \mu\text{m}$ , and  $5 \mu\text{m}$ , respectively.

agreement becomes poorer, as expected. Similar agreement is found for the TE waves and for higher order modes of both the TE and TM waves. Clearly, the simplicity of this model will aid us in figure analysis pertaining to mode conversion of a single incident wave.

#### IV. CONCLUSION

We have analyzed in detail the steady-state millimeter-wave propagation characteristics of Si and GaAs dielectric waveguides that contain a plasma-dominated region. We have calculated the phase shift and attenuation properties resulting from the presence of the plasma. Higher order modes were examined as well as frequency variation of the millimeter wave for all modes capable of propagating in a given sized guide. A surface plasma model was formulated that greatly simplified the analysis, yet gave good agreement with the more elaborate volume model. This model will facilitate the computation of a more complicated situation where mode coupling effects may be important. Phase shifts as high as  $1400^\circ/\text{cm}$  are predicted for modes near cutoff. This analysis indicates that in an oversized guide where many modes can be present, the properties of all modes are similar in both their expected phase shift as well as their attenuation properties.

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